Max SNR Array Processing for Space-Time Coded Systems

#### Hsuan-Jung Su

Department of Electrical Engineering National Taiwan University

# Outline

- Multi-Antenna Systems
- BLAST vs Space-Time Coding
- Rate-Performance Tradeoff: Grouped Space-Time Transmission
- Decoding of GST: Zero-Forcing vs Max SNR
- Performance
- Iterative Processing
- Conclusion

#### Multi-Antenna Systems



# **Channel Capacity**

• MIMO Channel Capacity (*n* Tx, *m* Rx)

 $C = \log \det[I_m + (\rho/n) \cdot HH^H]$ 

- *H* : Channel Gain Matrix
- $\rho$  : SNR
- Channel capacity increases linearly with min(*n*,*m*) when the parallel channels are uncorrelated with one another.
- In terms of practical use, the increased capacity amounts to enhanced power efficiency and/or spectral efficiency.

## Increased Capacity = ?

- Spectral Efficiency (Multiplexing Gain): Transmit as high data rate as possible but sacrifice power efficiency (e.g., BLAST).
- Power Efficiency (Diversity Gain):

Tx/Rx array processing, Tx coding and modulation design to achieve high spatial diversity (e.g., space-time coding). Each information bit spreads over many Tx antennas. Spectral efficiency is low.

### **BLAST** Transmission



## **BLAST** Reception



### **Space-Time Coding**



## **Space-Time Decoding**



 $P_e \leq a \cdot SNR^{-nm}$ 

Rate-Performance Tradeoff: Grouped Space-Time Transmission



For downlink in a cellular system, each group could be destined for one mobile.

### Decoding of GST



Sequential Decoding: Each stage consists of group interference suppression, decoding, and cancellation of the decoded group.

## **GST Signal Model**

$$r_t = Hc_t + n_t$$

$\boldsymbol{r}_{t} = \left[\boldsymbol{r}_{t}^{1}, \boldsymbol{r}_{t}^{2}, \boldsymbol{\Lambda}, \boldsymbol{r}_{t}^{m}\right]^{T}$			Received Signal		
$\boldsymbol{c}_{t} = \left[\boldsymbol{c}_{t}^{1}, \boldsymbol{c}_{t}^{2}, \boldsymbol{\Lambda}, \boldsymbol{c}_{t}^{n}\right]^{T}$			Simultaneously 7		
$\boldsymbol{n}_{t} = \left[\boldsymbol{n}_{t}^{1}, \boldsymbol{n}_{t}^{2}, \boldsymbol{\Lambda}, \boldsymbol{n}_{t}^{m}\right]^{T}$				AWGN	
H =	$\int h_{1,1}$	$h_{2,1}$	Λ	$h_{n,1}$	
	$h_{1,2}$	$h_{2,2}$	Λ	$h_{n,2}$	Ch
	M	Μ	Ο	M	CII
	$h_{1,m}$	$h_{2,m}$	Λ	$h_{n,m}$	

taneously Transmitted Symbols

**Channel Gain Matrix** 

## Group Interference Suppression I Zero-Forcing

When decoding group 1 (with  $n_1$  transmit antennas), use the null space of the interference space

$$\Lambda(C_{1}) = \begin{bmatrix} h_{n_{1}+1,1} & h_{n_{1}+2,1} & \Lambda & h_{n,1} \\ h_{n_{1}+1,2} & h_{n_{1}+2,2} & \Lambda & h_{n,2} \\ M & M & O & M \\ h_{n_{1}+1,m} & h_{n_{1}+2,m} & \Lambda & h_{n,m} \end{bmatrix}$$

as the array processor to remove completely signals from the other groups.

- Requirement:  $m \ge n n_1 + 1$
- Diversity:  $n_1(m-n+n_1)$ Transmit Receive

**Group Interference Suppression II** Max SNR Let  $R_{r}$  and  $R_{r}$  be the signal and interference covariance matrices, respectively, defined by  $R_{s} = H(C_{1})E[c_{t}^{1}(c_{t}^{1})^{H}]H^{H}(C_{1}), R_{a} = \Lambda(C_{1})E[c_{t}^{o}(c_{t}^{o})^{H}]\Lambda^{H}(C_{1}) + N_{0}I$ The linear maximum SNR array processor for this group consists of a set of k linearly independent eigenvectors corresponding to the non-zero eigenvalues, counting multiplicities, of the generalized eigenvalue problem  $R_{s}w = \lambda R_{p}w$ 

where  $k \leq \min(m, n_1)$  is the rank of  $R_s$ . The filtering outputs of these eigenvectors  $\{w_i^H r_t\}_{i=1}^k$  are uncorrelated with one another.



#### Zero-Forcing vs Max SNR

Assume perfect cancellation after decoding each group, when decoding group *i*:

	Zero-Forcing	Max SNR
Requirement	$m \ge n - \sum_{j=1}^{i} n_j + 1$	$m \ge 1$
Diversity	$n_i(m-n+\sum_{j=1}^i n_j)$	$\geq n_i(m-n+\sum_{j=1}^i n_j)$
Signal Energy	Partially Collected	Fully Collected
No. of Filters	$m-n+\sum_{j=1}^{i}n_{j}$	$\leq \min(m, n_i)$

#### Without Interfering Groups



#### Power Allocation (Open Loop)

- Zero-forcing diversity increases linearly with decoding stage → geometrically decreasing power allocation.
- Max SNR can use a more slowly decreasing power allocation (e.g., arithmetic).

#### **Spatial Interleaving**

The diversity gain after *group-based* spatial interleaving is no less than that provided by the space-time coding.



#### 4 Groups (3 Are Interfering)



#### Iterative Max SNR GST Receiver



#### **Performance Comparison**



## Conclusion

- Grouped space-time transmission achieves a tradeoff between transmission rate and performance.
- Max SNR array processing for GST outperforms ZF, and subsumes PRC. It requires a different power allocation than ZF to achieve a further performance gain.
- Max SNR array processing allows iterative processing. It can also be made adaptive and semi-blind for downlink applications where a mobile is aware of only its own group.